

PSET 4: Basic z-Transform

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1. Suppose $H(z)$ is such that it has a zero of multiplicity $m \geq 1$ at ∞ . Then it has an inverse transform such that $h[N] \neq 0$ but $h[n] = 0 \forall n \leq N - 1$. Find N .

This means $H(z)$ has a single causal region. We focus on this causal region. Multiplying it by z will decrease the multiplicity of the zero at ∞ and shift the corresponding impulse response (its inverse transform) $h[n]$ left once. When the multiplicity of the zero at ∞ becomes negative (i.e., when there is a pole at ∞), then there will be no causal region of $H(z)$ anymore.

The multiplicity of the zero at ∞ becomes zero when multiplying by z^m , which corresponds to $h[n + m]$. Since $h[n + m]$ is a causal signal and left-shifting it any more would make it non-causal by the above argument, this means that $h[m] \neq 0$ and $h[n] = 0 \forall n \leq m - 1$. Thus $N = m$.

If we take $m = 0$, that means H has a finite nonzero limit at $z \rightarrow \infty$. In this case, what can we conclude about the causal inverse transform $h[n]$?

With $N = m = 0$, $h[0] \neq 0$, by the above argument.

2. Given:

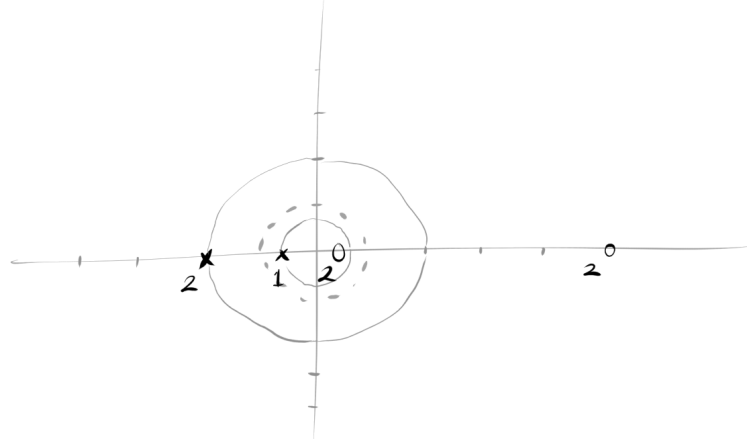
$$H(z) = \frac{3(2z - 1)^2(z - 5)^2}{(3z + 2)(z + 2)^2}$$

- (a) List all poles and zeros, with multiplicities, and draw a pole-zero plot by hand.

Poles: $-2/3$ ($m=1$), -2 ($m=2$), ∞ ($m=1$)

Zeros: $1/2$ ($m=2$), 5 ($m=2$)

Figure 1: Pole-zero plot



(b) Identify all possible ROCs for $H(z)$.

$$|z| < 2/3, 2/3 < |z| < 2, 2 < |z|$$

(c) Identify the ROC (if any) associated with:

- a causal system: none; pole at ∞
- a stable system: $2/3 < |z| < 2$
- a system with a well-defined frequency response: $2/3 < |z| < 2$

(d) We seek an integer L s.t. $z^L H(z)$ has a causal inverse transform such that $h[0] \neq 0$. Find all possible L .

This uses much of the same logic as question (1). For there to be a causal inverse transform, there must not be a pole at ∞ , so $L < 0$. The operation $z^L H(z)$ implies a left-shift of L on $h[n]$. Since $L < 0$ by the above statement, this means a right-shift by $-L$. Thus, specifying the condition $h[0] \neq 0$ puts a tight bound on L : we must shift $h[z]$ exactly enough units until it is causal, because only then will $h[0] \neq 0$ also be fulfilled.

By this reasoning, L must be -1 , since then $z^{-1}H(z)$ is the smallest right-shift that can occur to create a causal region.

3. A digital FIR filter with input x and output y is given by:

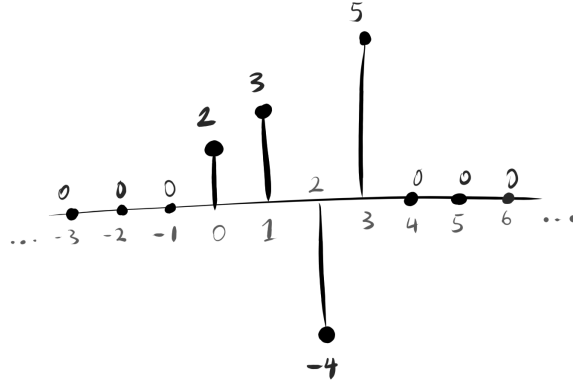
$$y[n] = 2x[n] + 3x[n-1] - 4x[n-2] + 5x[n-3]$$

(a) Sketch the impulse response $h[n]$. (See figure at top of next page.)

(b) Express $h[n]$ as a superposition of impulses.

$$h[n] = y[n](\delta) = 2\delta[n] + 3\delta[n-1] - 4\delta[n-2] + 5\delta[n-3]$$

Figure 2: Impulse response $h[n]$



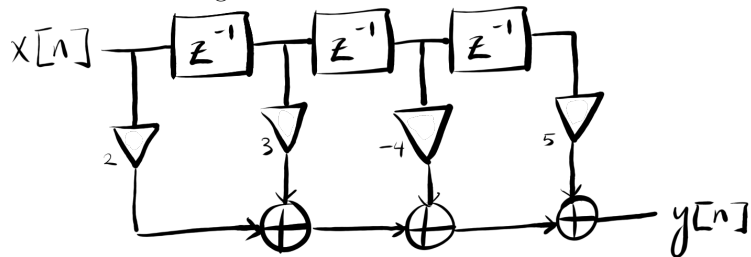
(c) Specify the length and order of the filter.

Length: 4 (number of terms)

Order: 3 (highest delay)

(d) Sketch a transversal filter realization.

Figure 3: Transversal realization.



4. Given the following transfer function of a digital IIR filter:

$$H(z) = \frac{3z^2 + 4z + 5}{10z^2 - z + 2}$$

(a) Write a difference equation with input x and output y .

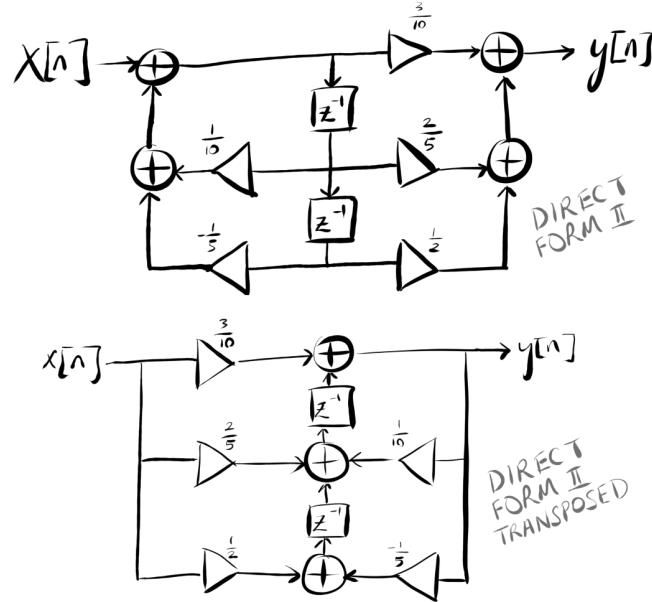
Rewriting in nicer form:

$$H(z) = \frac{\frac{3}{10}z^0 + \frac{2}{5}z^{-1} + \frac{1}{2}z^{-2}}{1 - (\frac{1}{10}z^{-1} - \frac{1}{5}z^{-2})}$$

$$y[n] = \frac{3}{10}x[n] + \frac{2}{5}x[n-1] + \frac{1}{2}x[n-2] + \frac{1}{10}y[n-1] - \frac{1}{5}y[n-2]$$

(b) Sketch direct form II and direct form II transposed realizations.

Figure 4: Direct form II and direct form II transposed realizations.



5. The following figure shows a block diagram. Find the overall transfer function by inspection.

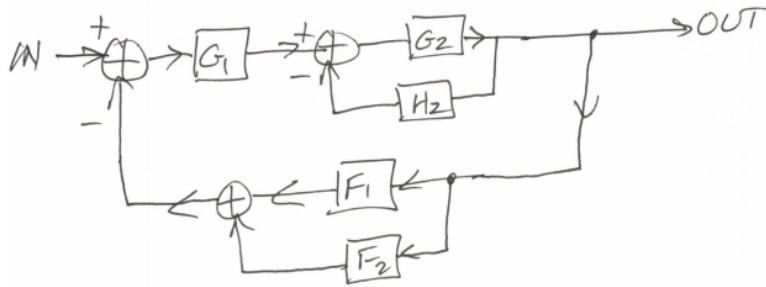


Figure 5: Block diagram of complex transfer function.

$$H = \frac{G_1 \left(\frac{G_2}{1+G_2H_2} \right)}{1 + G_1 \left(\frac{G_2}{1+G_1H_2} \right) (F_1 + F_2)}$$